

# ECE 754

## WARM-UP PROBLEM

GIVEN:  $w[n] = \begin{cases} 1 & 0 \leq n \leq N-1 \\ 0 & \text{OTHER} \end{cases}$

FIND DTFT OF  $w[n]$  + SKETCH  $|W(e^{j\omega})|$

FIND z-TRANS. OF  $w[n]$

SKETCH POLES + ZEROS OF  $W(z)$

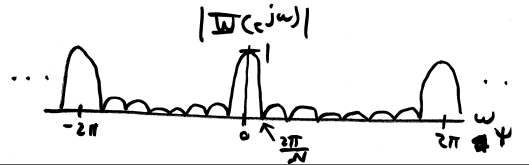
## WARM-UP

$$W(z) = \sum_n w[n] z^{-n} = \frac{1-z^{-N}}{1-z^{-1}}$$

$$W(e^{j\omega}) = \sum_n w[n] e^{-j\omega n} = \frac{1-e^{-j\omega N}}{1-e^{-j\omega}}$$

How DO z + FOURIER RELATE?

$$W(e^{j\omega}) = W(z) \Big|_{z=e^{j\omega}} \quad z \text{ ON UNIT CIRCLE}$$

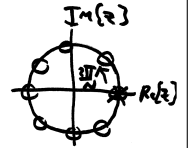


## WARM-UP

$$W(z) = \frac{1-z^{-N}}{1-z^{-1}} \cdot \frac{z^N}{z^N} \quad \text{POLES + ZEROS?}$$

$$= \frac{z^N - 1}{z^{N-1}(z-1)}$$

ZEROS:  $z^N = 1 = e^{j2\pi n}$   
 $z = e^{j\frac{2\pi n}{N}}$



POLES:  $z^{N-1} = 0$   
 $z = 0$  N-1 POLES AT  $z=0$   
 1 POLE AT  $z=1$

## CH. 3 : DESIGN PROBLEM

CHOOSE  $w$  TO MEET CERTAIN CRITERIA

EX) MINIMIZE BEAMWIDTH (FILTER PASSBAND WIDTH)

CONTROL SIDELobe LEVELS

UNCERTAINTY PRINCIPLE:

$$\sqrt{\Delta x^2} \sqrt{\Delta k^2} \geq \frac{1}{2}$$

NORMALIZED MEAN SQ. WIDTH OF APERTURE

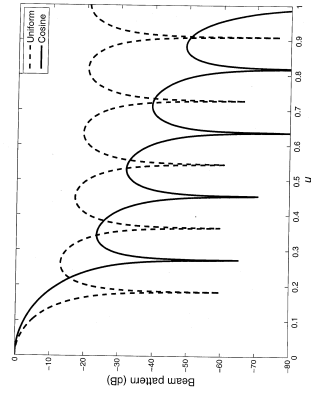
NORM. MEAN SQ WIDTH OF SPATIAL RESPONSE

## FOCUS ON LINEAR, EQUALLY SPACED ARRAYS

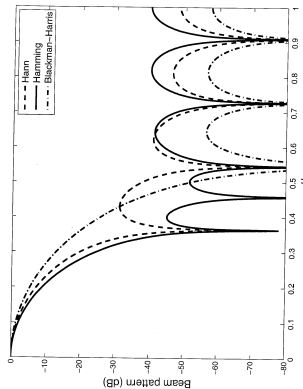
DESIGN METHODS:

- $w$  = STANDARD WINDOW AS USED IN SPECTRAL ESTIMATION
- FREQUENCY SAMPLING
- MINIMUM BW FOR FIXED SIDELobe LEVEL → KOLPH-CHEBYCHEV
- LEAST SQUARES  
MINIMIZE MSQ ERROR BETWEEN DESIRED + ACTUAL BEAM PATTERN
- MIN-MAX (PARKS-McCLELLAN)

3.1 OF TEXT



TRADEOFF: M.L. WIDTH SIDELobe HEIGHT



## On the Use of Windows for Harmonic Analysis with the Discrete Fourier Transform

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**I. INTRODUCTION**

By recently, every observed signal we process must be of finite duration. This is true for all signals and systems, but it is particularly true for the signals and systems that we use in the analysis of signals and systems. The signals and systems that we use in the analysis of signals and systems are of finite duration because they are of finite duration in the time domain. The signals and systems that we use in the analysis of signals and systems are of finite duration because they are of finite duration in the time domain.

**II. HARMONIC ANALYSIS OF FINITE-DURATION SIGNALS**

We observe that by defining  $x[n]$  as an observed signal, we are defining the spectrum over a finite interval. The spectrum of a signal is defined over a finite interval because the signal is of finite duration. The spectrum of a signal is defined over a finite interval because the signal is of finite duration.

## FREQ. - SAMPLING FOR UNIFORMLY - SPACED ARRAYS

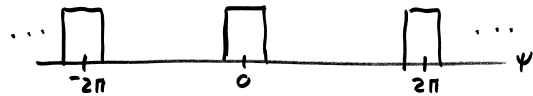
RECALL:  $B(\psi) = \mathcal{F}\{\psi\} = e^{-j(\frac{N-1}{2})\psi} \left( \sum_{n=0}^{N-1} w[n] e^{-jn\psi} \right) e^{-jn\psi}$

DFT OF  $w[n]$  (WEIGHT SEQUENCE)

ARRAY CENTERED AT  $z=0$

REARRANGE:  $B^*(\psi) e^{-j(\frac{N-1}{2})\psi} = \sum_{n=0}^{N-1} w[n] e^{-jn\psi}$

DTFT'S ARE PERIODIC w/ PERIOD  $2\pi$   
DTFT



CAN WE GET  $w_n$  BY SAMPLING THIS DESIRED DTFT? YES  $\rightarrow$  DFT

DFT: SAMPLES EQUALLY SPACED 0 TO  $2\pi$

VAN TREES:  $N$  EQUALLY SPACED SAMPLES FROM  $-\pi$  TO  $+\pi$

$$\psi_k = \left(k - \left(\frac{N-1}{2}\right)\right) \frac{2\pi}{N} \quad k=0, \dots, N-1$$

$$B(k) = B^*(\psi_k) e^{-j\left(\frac{N-1}{2}\right)\psi_k} = \sum_{n=0}^{N-1} w_n e^{-jn\psi_k}$$

$$B(k) = \sum_{n=0}^{N-1} w_n e^{-jn\left(k - \left(\frac{N-1}{2}\right)\frac{2\pi}{N}\right)}$$

$$= \sum_{n=0}^{N-1} w_n e^{+jn\left(\frac{N-1}{2}\right)\pi} e^{-jnk\frac{2\pi}{N}}$$

$$B(k) = \sum_{n=0}^{N-1} \underbrace{w_n e^{+jn\left(\frac{N-1}{2}\right)\pi}}_{b_n} e^{-j\frac{2\pi}{N}nk} \leftarrow \text{LOOKS LIKE DFT!}$$

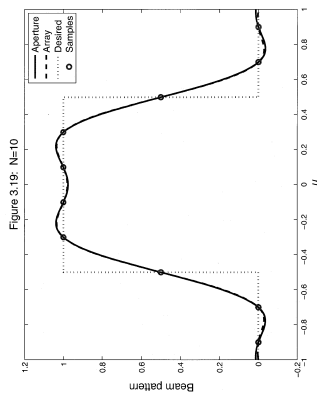
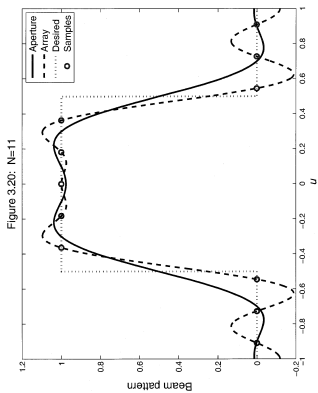
GIVEN:  $B(k)$  SAMPLES OF DESIRED PATTERN

WANT:  $b_n \rightarrow w_n = b_n e^{-jn\left(\frac{N-1}{2}\right)\pi}$

$$\text{IDFT: } b_n = \frac{1}{N} \sum_{k=0}^{N-1} B(k) e^{+j\frac{2\pi}{N}nk}$$

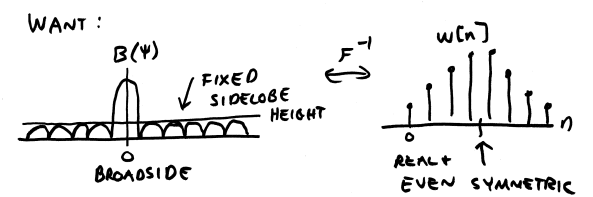
WILL GET  $B(\psi) = B_{\text{DESIRED}}(\psi) \left| \psi = \psi_k \right.$   
SAMPLE POINT

COMPUTE  $B(\psi) = \underline{w}^H \underline{v}(\psi)$  FOR ALL  $\psi$  TO GET PERFORMANCE



MINIMUM BEAMWIDTH FOR SPECIFIED SIDELobe LEVEL

DOLPH-CHEBYCHEV APPROACH



$$B(\psi) = \underline{w}^H \underline{v}(\psi) = \sum_{n=0}^{N-1} w_n^* e^{j\psi\left(n - \left(\frac{N-1}{2}\right)\right)}$$

IF  $N = \text{ODD}$

$$B(\psi) = \sum_{m=-\left(\frac{N-1}{2}\right)}^{+\left(\frac{N-1}{2}\right)} a_m e^{j\psi m} \quad a_m = w_n \Big|_{n=M+\frac{N-1}{2}}$$

ASSUMING  $a_m$  REAL + EVEN

$$B(\psi) = a_0 + 2 \sum_{m=1}^{\frac{N-1}{2}} a_m \cos(m\psi)$$

IF  $N = \text{EVEN}$

$$B(\psi) = 2 \sum_{m=1}^{N/2} a_m \cos\left(\left(m - \frac{1}{2}\right)\psi\right)$$

REWRITING:

ODD  $B(\psi) = \sum_{k=0}^{\frac{N-1}{2}} \alpha_k \cos\left(2k \frac{\psi}{2}\right)$

EVEN  $B(\psi) = \sum_{k=1}^{N/2} \alpha_k \cos\left((2k-1) \frac{\psi}{2}\right)$

$$\alpha_k = \begin{cases} 2\alpha_k & k \neq 0 \\ \alpha_k & k = 0 \end{cases}$$

DOLPH-CHEBYCHEV:  
 $B(\psi)$  AS POLYNOMIAL IN  $\cos(\psi/2)$  OF ORDER  $N-1$

LOOK AT TERM LIKE  $\cos\left(M \frac{\psi}{2}\right)$

$$\cos\left(M \frac{\psi}{2}\right) = \text{Re} \left\{ \left( e^{j\frac{\psi}{2}} \right)^M \right\}$$

$$= \text{Re} \left\{ \left[ \cos\left(\frac{\psi}{2}\right) + j \sin\left(\frac{\psi}{2}\right) \right]^M \right\}$$

BINOMIAL THM:

$$(a+b)^M = \sum_{l=0}^M \binom{M}{l} a^{M-l} b^l$$

$$\binom{M}{l} = \frac{M!}{l!(M-l)!}$$

EX |  $M=2$

$$\begin{aligned} \cos\left(2\left(\frac{\psi}{2}\right)\right) &= \operatorname{Re}\left\{\left[\cos\left(\frac{\psi}{2}\right) + j\sin\left(\frac{\psi}{2}\right)\right]^2\right\} \\ &= \operatorname{Re}\left\{\cos^2\left(\frac{\psi}{2}\right) + 2j\cos\left(\frac{\psi}{2}\right)\sin\left(\frac{\psi}{2}\right) - \sin^2\left(\frac{\psi}{2}\right)\right\} \\ &= \cos^2\left(\frac{\psi}{2}\right) - \sin^2\left(\frac{\psi}{2}\right) \\ &= \cos^2\left(\frac{\psi}{2}\right) - (1 - \cos^2\left(\frac{\psi}{2}\right)) \\ &= 2\cos^2\left(\frac{\psi}{2}\right) - 1 = \cos\left(2\left(\frac{\psi}{2}\right)\right) \end{aligned}$$

$B(\psi) = N-1$  ORDER POLYNOMIAL  
IN POWERS OF  $\cos\left(\frac{\psi}{2}\right)$

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$$\cos\left(M\frac{\psi}{2}\right) = T_M(x) \Big|_{x=\cos\left(\frac{\psi}{2}\right)}$$

$T_M \triangleq$  MTH CHEBYCHEV POLYNOMIAL

$$T_0(x) = 1$$

$$T_1(x) = x$$

$$T_2(x) = 2x^2 - 1$$

$$T_3(x) = 4x^3 - 3x$$

⋮

RECURSION:  
 $M \geq 2$

$$T_M(x) = 2xT_{M-1}(x) - T_{M-2}(x)$$

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$$T_M(x) = \begin{cases} \cos(M\cosh^{-1}(x)) & |x| \leq 1 \\ \cosh(M\cosh^{-1}(x)) & x > 1 \\ (-1)^M \cosh(M\cosh^{-1}(x)) & x < -1 \end{cases}$$

PROPERTIES:

\*  $T_M(x)$  HAS  $M$  REAL ROOTS IN  
INTERVAL  $|x| < 1$

$$T_M(x) = \cos\left(M\frac{\psi}{2}\right) \Big|_{\cos\left(\frac{\psi}{2}\right) = x}$$

$$\text{ROOTS: } \cos\left(M\frac{\psi}{2}\right) = 0$$

$$M\frac{\psi}{2} = \frac{\pi}{2}(2p-1) \quad p=1, \dots, M$$

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ROOTS IN  $\psi$  SPACE:

$$\psi = \pi \left(\frac{2p-1}{M}\right)$$

IN  $x$ -SPACE ROOTS ARE:

$$x_p = \cos\left((2p-1)\frac{\pi}{2M}\right)$$

PROPERTIES:

\*  $T_M(x)$  HAS ALTERNATING MINIMA +  
MAXIMA

$$|T_{\min}| = |T_{\max}| = 1$$

EQUIRIPPLE IN  $x$  VARIABLE

\* AT  $x = \pm 1$ ,  $|T_M(x)|_{x=\pm 1} = 1$   
 $x > 1$   $|T_M(x)| > 1$

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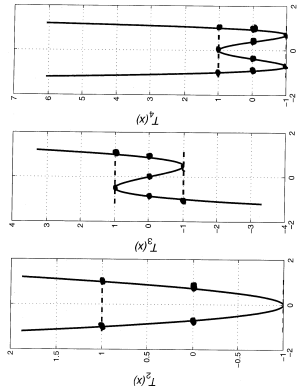


Figure 3-23

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DOLPH-CHEBY IDEA:

\*  $B(\psi)$  IS AN  $N-1$  ORDER POLYNOMIAL IN  
 $\cos\left(\frac{\psi}{2}\right)$

\* SET  $B(\psi) =$  CHEBYCHEV POLYNOMIAL  
SO WE GET EQUIRIPPLE PROPERTY  
DEFINE A MAPPING SO THAT  $B(\psi)$  MAXIMUM  
(M.L. MAX) CORRESPONDS TO A PT  $x_0 > 1$

$$R = \frac{\text{M.L. MAX}}{\text{SIDELobe LEVEL}}$$

SOLVE FOR  $x_0$ :

$$T_{N-1}(x_0) = R = \cosh\left((N-1)\cosh^{-1}(x_0)\right)$$

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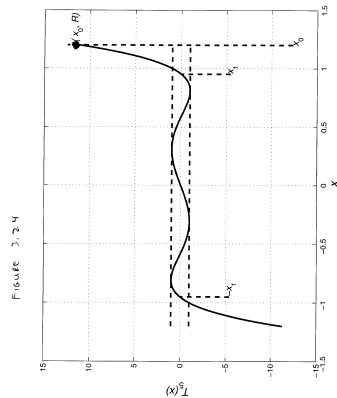


Figure 3-24

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$$x_0 = \cosh\left(\frac{1}{N-1}\cosh^{-1}(R)\right)$$

RECALL MAPPING:

$$x = \cos\left(\frac{\psi}{2}\right)$$

RESTRICTS  $-1 \leq x \leq +1$

ADJUST MAPPING:

$$q = \frac{x}{x_0} = \cos\left(\frac{\psi}{2}\right) \quad (q = w \text{ IN TEXTBOOK})$$

$$x = x_0 \cos\left(\frac{\psi}{2}\right)$$

$$\text{BEAMPATTERN: } B(\psi) = \frac{1}{R} T_{N-1}\left(x_0 \cos\left(\frac{\psi}{2}\right)\right)$$

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HOW TO FIND WEIGHTS GIVEN  $B(\psi) =$  CHEBY  
POLY.

$$\underline{w}^H \underline{v}(\psi) \Big|_{\psi=0} = 1 \quad \text{MAINLOBE CONSTRAINT}$$

$$\underline{w}^H \underline{v}(\psi_p) = 0 \quad N-1 \text{ ZEROS} \\ (\text{NULLS IN BEAMPATTERN})$$

MATRIX FORM:

$$\underline{w}^H \underline{V} = [1 \ 0 \ 0 \ \dots \ 0]$$

$$\text{OR } \underline{V}^H \underline{w} = \begin{bmatrix} 1 \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix} \quad \underline{V} = \begin{bmatrix} 1 & 1 & \dots & 1 \\ v(0) & v(\psi_1) & \dots & v(\psi_p) \\ 1 & 1 & \dots & 1 \end{bmatrix}$$

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SOLVING:

$$\underline{w} = (V^H)^{-1} \begin{bmatrix} 1 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$

TEXT CALLS THIS  $\underline{e}_1$

KNOW: WHERE ARE  $\Psi_p$ 's? NULLS

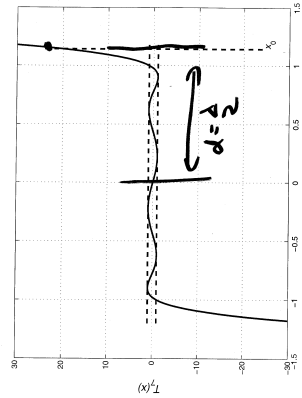
ORIGINAL SPACE:  $\Psi = \frac{\pi}{N-1} (2p-1)$   $p=1, \dots, M$

$$x_p = \cos\left(\frac{\pi}{2} \left(\frac{2p-1}{N-1}\right)\right)$$

NEW SPACE:  $q_p = \frac{x}{x_0} = \frac{1}{x_0} \cos\left(\frac{\pi}{2} \left(\frac{2p-1}{N-1}\right)\right)$

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FIGURE 3.25



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NEW ZEROS:  $\Psi_p = 2 \cos^{-1}(q_p)$

$$\Psi_p = 2 \cos^{-1}\left(\frac{1}{x_0} \cos\left(\frac{\pi}{2} \left(\frac{2p-1}{N-1}\right)\right)\right) \quad p=1, \dots, N-1$$

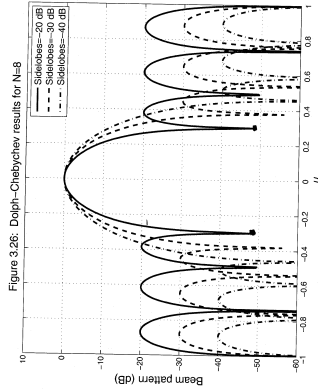
MAPPING:

$$B(\Psi) = \frac{1}{R} T_{N-1}\left(x_0 \cos\left(\frac{\Psi}{2}\right)\right) = T_{N-1}(x)$$

$$x = x_0 \cos\left(\frac{\Psi}{2}\right)$$

$\theta$	0	$\pi/2$	$\pi$
$\Psi = \frac{2\pi}{\lambda} d \cos \theta$	$\frac{2\pi d}{\lambda}$	0	$-\frac{2\pi d}{\lambda}$
$x = x_0 \cos\left(\frac{\Psi}{2}\right)$	$x_0 \cos\left(\frac{\pi d}{\lambda}\right)$	$x_0$	$x_0 \cos\left(\frac{\pi d}{\lambda}\right)$

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MAPPING:

$$d = \frac{\lambda}{4} : x_0 \cos\left(\frac{\pi \lambda / 4}{\lambda}\right) \leq x \leq x_0$$

$$x_0 \cos\left(\frac{\pi}{4}\right) \leq x \leq x_0$$

$$.707 x_0 \leq x \leq x_0$$

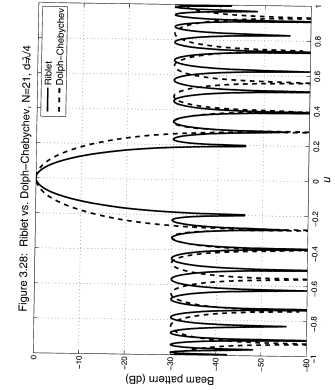
$$d = \frac{\lambda}{2} : x_0 \cos\left(\frac{\pi \lambda / 2}{\lambda}\right) \leq x \leq x_0$$

$$0 \leq x \leq x_0$$

$$d = \frac{3\lambda}{4} : -.707 x_0 \leq x \leq x_0$$

$$d \leq \frac{\lambda}{\pi} \cos^{-1}\left(\frac{-1}{x_0}\right) \leftarrow \text{FOR MAPPING TO BE VALID (LEFT EDGE = -1)}$$

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### RIBLET - CHEBYCHEV

- USES EXPANSION IN TERMS OF  $\cos(\Psi)$  INSTEAD  $\cos\left(\frac{\Psi}{2}\right)$
- POLYNOMIAL OF ORDER  $\frac{N-1}{2}$  (N ODD)
- MAPPING:  $x = c_1 \cos \Psi + c_2$   
SOLVE FOR  $c_1 + c_2$  SO THAT FULL RANGE OF CHEBYCHEV POLYNOMIAL IS USED

$$-1 \leq x \leq x_0$$

$$d = \frac{\lambda}{2} \text{ NO DIFFERENCE} \quad \left| \quad d < \frac{\lambda}{2} \text{ DIFFERENCE} \right.$$

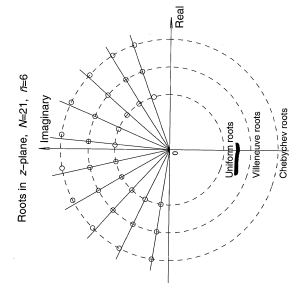
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### VILLENEUVE

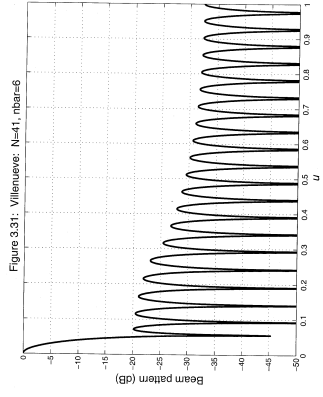
- START W/ ZEROS ASSOCIATED W/ UNIFORM WEIGHTING (WARM-UP)
  - EXCHANGE  $2\bar{n}$  OF THOSE NEAR BROADSIDE FOR  $\approx$  CHEBYCHEV ROOTS
- $\Rightarrow$  RESULT: BETTER S.L. NEAR BROADSIDE  
DECREASING S.L. FARTHER OUT

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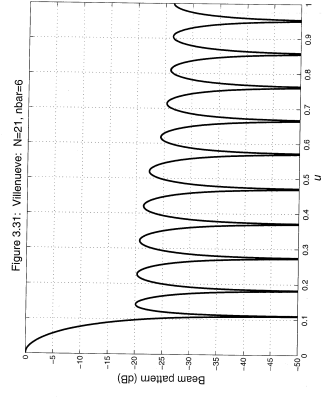
FIGURE 3.30



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